

# A Hierarchical Test Generation Technique for Embedded Systems

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**Abstract**<sup>1</sup> *This paper presents a hierarchical test generation technique for embedded systems containing hardware and software. The technique is applied to the system-level specification of such systems. Different from the traditional approaches, hardware and software parts of an embedded system are handled in a uniform way. We will in particular show how the proposed technique can be applied at high levels of abstraction and how the software domain of the specification can also be successfully covered.*

## 1. Introduction

The development of hardware/software codesign techniques has made it possible to design hardware and software of an embedded system at the high levels of abstraction in a uniform way. However the testing of the hardware and software parts of the system are still considered as totally unrelated problems and solved with very different methods.

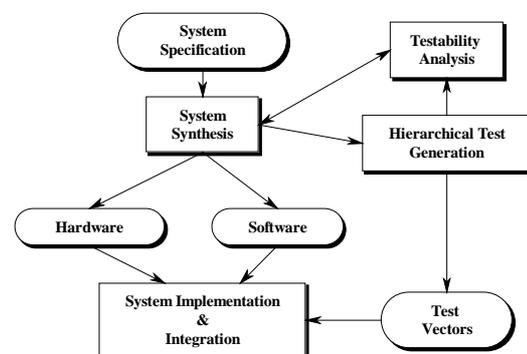
In the early phases of the design cycle, system synthesis is performed starting from an implementation independent specification. Reasoning about testability at this level can be facilitated by an uniform test generation technique, which is both applicable to the hardware and software domains. In [1], [4-5], [13] test generation and testability analysis for this particular problem has been studied, but not many efficient techniques have been developed yet.

In our approach, testability evaluation and test generation at the system level are based on hierarchical test generation (HTG) [9]. We apply HTG, using a decision diagram (DD) [12] based representation, and show that it can be used for both the hardware and software domains as well as for different levels of abstraction.

## 2. Hierarchical Test Generation for Hardware/Software Systems

Test generation has been proven to be an NP-complete problem [6]. There has been a lot of research devoted to solve the test generation problem for gate-level circuits. Working at this level provides very high quality of the tests but is computationally very expensive in the case of large circuits and therefore practically not usable. Several approaches have been developed to handle test generation for relatively large combinational circuits in a reasonable time. Test generation for large sequential circuits remains, however, an unsolved problem, despite rapid increase of computational power. Hierarchical test generation has been proposed as one possible solution [2, 8, 10].

To give the designer an opportunity to perform design for testability already in the



**Figure 1. Test generation and testability analysis in a hardware/software co-design environment**

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early design stages, testability evaluation should be applied directly to the system specification. And a testability metric should be part of the cost function considered during system level synthesis, and in particular for hardware/software partitioning.

Figure 1 shows how testability evaluation and test generation fit into such a system synthesis concept. As discussed in section 1, testability evaluation and test generation are performed on a high level implementation-independent representation and they provide results to be interpreted and used in a coherent way for both the hardware and software partitions.

### 3. HTG for Specifications to be Implemented in Software

In our approach, decision diagrams are used for design modeling at the high abstraction levels. The main advantage of modelling with DDs lies in the fact that a uniform concept can be applied on different abstraction levels. An extended overview of DDs is presented in [12].

Our main objectives are to show how DDs can be used for test generation at the behavioural level and how HTG can be used for testing the part of the system which finally will be implemented as software. Hierarchical test generation technique for hardware has been reported at [8].

At this level, for every internal variable and primary output of the design a data-flow DD will be generated. Terminal nodes of the data-flow DD represent arithmetic expressions. Further, an additional DD which describes the control-flow has to be generated. The control-flow DD describes the succession of statements and branch activation conditions. Figure 2 depicts an example of DD, describing the behavior of a simple function. For example, variable A will be equal to  $INI+2$ , if the system is in the state  $q=2$  (Figure 2c). If this state is to be activated, condition  $INI \geq 0$  should be true (Figure 2b). The DDs extracted from a specification will be used as a computational model in HTG for symbolic path activation.

#### 3.1. Test Generation Algorithm

There are two types of tests which we consider in the current approach. One set targets nonterminal nodes of the control-flow DD (conditions for branch activation) and the second set aims at testing operators, depicted in terminal nodes of the data-flow DD.

The whole test generation task is performed in the following way. Tests are generated sequentially for each nonterminal node of the control-flow DD. Symbolic path activation is performed and functional constraints are extracted. Solving the constraints gives us the path activation conditions to reach a particular segment of the specification. In order to test the operations, presented in the terminal nodes of the data-flow DD, different approaches can be

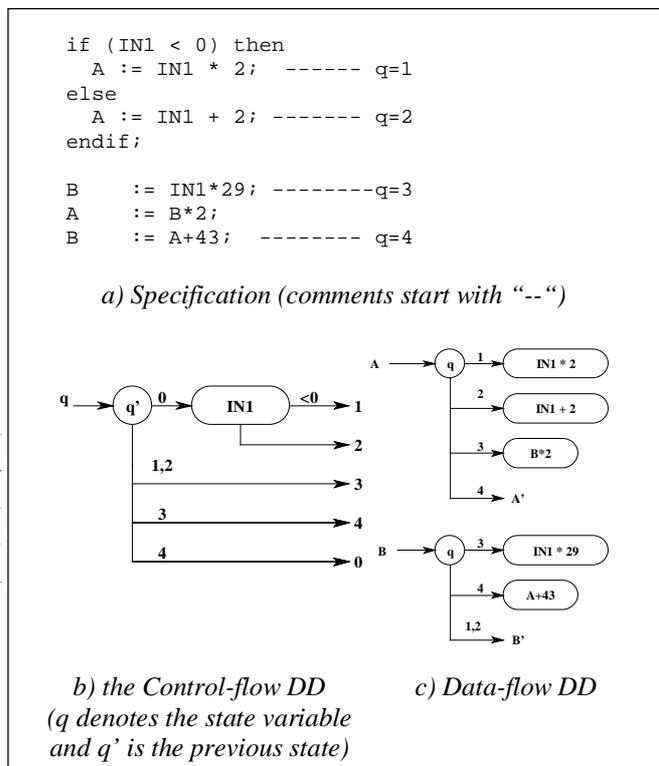


Figure 2. A DD example

used. In this paper, we use mutation testing [4] for test generation for the operations at the terminal nodes. For path activation, a slightly modified version of the algorithm described in [8] is used.

### 3.2. Conformity Test

For the nonterminal nodes of the control-flow DD, conformity tests will be applied. The conformity tests target errors in branch activation. In order to test nonterminal node  $INI$  (Figure 3), one of the output branches of this node should be activated. Activation of the output branch means activation of a certain set of program statements. In our example, activation of the branch  $INI < 0$  will activate the branches in the data-flow DD where  $q=1$  ( $A:=X$ ). For observability the values of the variables calculated in all the other branches of  $INI$  have to be distinguished from the value of the variables calculated by the activated branch. In our example, node  $INI$  is tested, in the case of  $INI < 0$ , if  $X \neq Y$ . The path from the root node of the control-flow DD to the node  $INI$  has to be activated to ensure the execution of this particular specification segment. The conditions, generated here, should be justified to the primary inputs of the module. This process will be repeated for each output branch of the node. In the general case there will be  $n(n-1)$  tests, for every node, where  $n$  is the number of output branches.

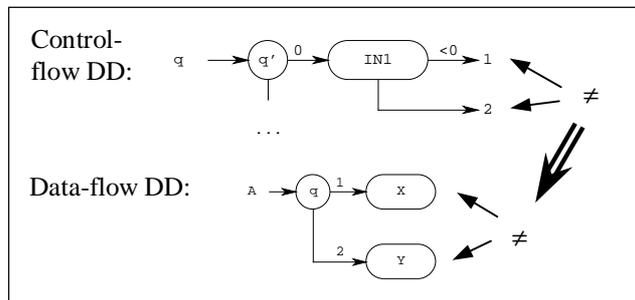


Figure 3. Conformity test

### 3.3. Testing Arithmetic Operators

As mentioned earlier, test vectors for the terminal nodes can be generated based on different approaches, and our HTG technique does not impose a specific one. Currently we use a mutation based fault model [4] for testing terminal nodes of the data-flow DD. We are using a library of operator mutations, which describes for each operator a set of corresponding mutants and conditions, which can distinguish between the mutant and the original operator. Suppose we have the expression:  $x := (a+b) - c$ . To rule out the fault that the first “+” is changed to “-”,  $b$  must not be 0 (because  $a+0=a-0$ ). Additionally, to rule out the fault that instead of “+” there is “\*”, we have to assure that  $a+b \neq a*b$ . For more details about operator mutants, we refer the reader to [7].

## 4. Experimental Results

Experiments were conducted in the environment consisting of our hierarchical test generator, the library of mutants for different arithmetic operators, and the Generic Coverage Tool (GCT) [11] which measures the quality of the generated test cases. Conversion between different representations (VHDL, C, Fortran and DD) is performed by corresponding translation tools. In order to evaluate our results we compare them with those produced by the software test generation tool Mothra [3]. Experiments were carried out on three embedded software examples. Table 1 presents the experimental results of our approach in comparison with the results achieved by Mothra. The Achieved fault coverage reflects synthetically several different coverage criteria (statement coverage, branch coverage, loop coverage etc.). As observed, the mutation based testing tool Mothra generates a much larger set of test vectors, which, at the same time, produces a weaker coverage.

Design module	Number of lines in the specification	Number of branches	Number of mutants	Mothra			Our approach	
				Number of generated test cases	Number of optimized test cases	Fault coverage	Number of generated test cases	Fault coverage
Square	38	12	813	707	5	77.65%	10	94.12%
Mult	20	6	478	449	3	84.00%	6	90.00%
FFT	31	4	1682	1639	4	83.91%	6	86.21%

**Table 1. Experimental results**

## 5. Conclusions

This paper describes a novel hierarchical framework for test generation in hardware/software systems. Hardware and software parts of an embedded system can be handled in a uniform way. The same DD representation can be used for describing systems at different abstraction levels, including the system level. Based on this representation, reasoning about testability in the early design phases and test generation for both the hardware and the software domain is possible. We have shown how HTG can be applied at the high levels of abstraction and how the software domain of the specification can be successfully covered. Experimental results have shown that the quality of the generated test vectors is even better than those produced by specific software test generation tools. Our future research is to integrate the testability metrics based on HTG into a hardware/software codesign environment.

## 6. Acknowledgements

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